## UNIVERSITY OF BOLTON

## SCHOOL OF ENGINEERING

## BENG (HONS) ELECTRICAL \& ELECTRONICS ENGINEERING

## SEMESTER ONE EXAMINATION 2018/2019

## ENGINEERING ELECTROMAGNETISM

## MODULE NO: EEE6012

Date: Wednesday $16^{\text {th }}$ January 2019 Time: 14:00-16:00

INSTRUCTIONS TO CANDIDATES:
There are SIX questions.
Answer ANY FOUR questions.
All questions carry equal marks.

Marks for parts of questions are shown in brackets.

Electronic calculators may be used provided that data and program storage memory is cleared prior to the examination.

CANDIDATES REQUIRE:
Formula Sheet (attached).

School of Engineering
BEng (Hons) Electrical \& Electronic Engineering
Semester One Examination 2018/2019
Engineering Electromagnetism
Module no. EEE6012

## Question 1

A) Three symmetrical traveling waves $X, Y$, and $Z$ have the following characteristics:

$$
v_{X}=5 \cos 4 \pi x \text { Volts. }
$$

| Waveform | First peak <br> value(Volts) at <br> $\mathrm{x}=0$ metre | Second peak <br> value (Volts) at <br> $\mathrm{x}=0.5$ metre |
| :--- | :---: | :---: |
| Waveform X | 5 | 5 |
| Waveform Y | 5 | 3.52 |
| Waveform Z | 5 | 1.01 |

What expression is applicable to
i) the $Y$ wave?
ii) the $Z$ wave?
B) An electromagnetic wave is propagating in the z-direction in a lossy medium with attenuation constant $\alpha=0.45 \mathrm{~Np} / \mathrm{m}$. If the wave's electric-field amplitude is 120 $\mathrm{V} / \mathrm{m}$ at $\mathrm{z}=0$, how far can the wave travel before its amplitude will have been reduced to
i) $12 \mathrm{~V} / \mathrm{m}$
ii) $1.2 \mathrm{~V} / \mathrm{m}$
iii) $1.2 \mu \mathrm{~V} / \mathrm{m}$
C) An RL series circuit is connected to a single phase supply with instantaneous voltage of $v_{s}=20 \sin \left(4 \times 10^{4} t-30^{\circ}\right)$ Volts. If the resistance is $3 \Omega$ and the inductance is 0.1 mH . Find:
i) $v_{s}$ in cosine form
ii) the current I in polar form
iii) the voltage across the inductance in phasor and time domains

School of Engineering
BEng (Hons) Electrical \& Electronic Engineering
Semester One Examination 2018/2019
Engineering Electromagnetism
Module no. EEE6012

## Question 2

A) Find the distance vector between $\mathbf{P}_{\mathbf{1}}(2,2,-3)$ and $\mathbf{P}_{\mathbf{2}}(-1,-2,3)$ in Cartesian coordinates and its magnitude in centimetres.
B) Find the angle $\theta$ between vectors $\mathbf{L}$ and $\mathbf{M}$ of branch $A$ using the cross product between them.
[12 marks]
C) Find the angle that vector $\mathbf{M}$ of branch $A$ makes with the $z$-axis.
[6 marks]

## Total 25 marks

## Question 3

A) Four charges of $20 \mu \mathrm{C}$ each are located in free space at points with Cartesian coordinates ( $-4,0,0$ ), ( $4,0,0$ ), ( $0,-2,0$ ), and ( $0,4,0$ ). Find the force on a $40 \mu \mathrm{C}$ charge located at ( $0,0,3$ ). All distances are in metres.
[12 marks]
B) A wire is formed into a square loop and placed in the $x-y$ plane with its centre at the origin and each of its sides parallel to either the $x$ - or $y$-axes. Each side is 30 cm in length, and the wire carries a current of 10 Amperes whose direction is clockwise when the loop is viewed from above. Calculate the magnetic field at the centre of the loop.
[13 marks]

Total 25 marks

School of Engineering
BEng (Hons) Electrical \& Electronic Engineering
Semester One Examination 2018/2019
Engineering Electromagnetism
Module no. EEE6012

## Question 4

a) A $10-\mathrm{MHz}$ uniform plane wave is traveling in a nonmagnetic medium with $\mu=\mu_{0}$ and $\varepsilon r=9$. Find the following parameters:
(i) the phase velocity,
(ii) the wavenumber,
(iii) the wavelength in the medium,
[3 marks]
(iv) the intrinsic impedance of the medium.
b) The electric field phasor of a uniform plane wave traveling in a lossless medium with an intrinsic impedance of $188.5 \Omega$ is given by $\tilde{E}=10 e^{-j 4 \pi y} \hat{Z}(\mathrm{mV} / \mathrm{m})$.
(i) Determine the associated magnetic field phasor
(ii) Find the instantaneous expression for $E(y, t)$ if the medium is nonmagnetic ( $\mu=\mu_{0}$ ).
[8 marks]

Total 25 marks

PLEASE TURN THE PAGE....

School of Engineering
BEng (Hons) Electrical \& Electronic Engineering
Semester One Examination 2018/2019
Engineering Electromagnetism
Module no. EEE6012

## Question 5

(a) A telephone line has the following parameters:
$\mathrm{R}=40 \Omega / \mathrm{m}, \mathrm{G}=400 \mu \mathrm{~S} / \mathrm{m}, \mathrm{L}=0.2 \mu \mathrm{H} / \mathrm{m}, \mathrm{C}=0.5 \mathrm{nF} / \mathrm{m}$
(i) If the line operates at 10 MHz , calculate the characteristic impedance Zo and velocity $u$.
(ii) After how many metres will the voltage drop by 30 dB in the line?
(b) Fig.5(b) shows a lossless transmission line model.


Fig.5(b) lossless transmission line model
(i) Find reflection coefficient $\Gamma$ and standing wave ratio $S$
(ii) Determine $Z_{\text {in }}$ at the generator.

School of Engineering
BEng (Hons) Electrical \& Electronic Engineering
Semester One Examination 2018/2019
Engineering Electromagnetism
Module no. EEE6012

## Question 6

(a) An antenna has a conical radiation pattern with a normalized radiation intensity $F(\theta)=1$ for $\theta$ between $0^{\circ}$ and $45^{\circ}$ and zero intensity for $\theta$ between $45^{\circ}$ and $180^{\circ}$. The pattern is independent of the azimuth angle $\phi$.
(i) Find the pattern solid angle
[5 marks]
(ii) Determine the directivity
[2 marks]
(b) The maximum power density radiated by a short dipole at a distance of 1 km is $60(\mathrm{nW} / \mathrm{m} 2)$. If $\mathrm{IO}=10 \mathrm{~A}$, find the radiation resistance.
[8 marks]
(c) A 3-GHz microwave link consists of two identical antennas each with a gain of 30 dB . Determine the received power, given that the transmitter output power is 1 kW and the two antennas are 10 km apart.
(d) The effective area of an antenna is 9 m 2 . What is its directivity in decibels at 3 GHz ?

## END OF QUESTIONS

Formula sheet over the page....

School of Engineering
BEng (Hons) Electrical \& Electronic Engineering
Semester One Examination 2018/2019
Engineering Electromagnetism
Module no. EEE6012

## Formula sheet

These equations are given to save short-term memorisation of details of derived equations and are given without any explanation or definition of symbols; the student is expected to know the meanings and usage.

Time-domain sinusoidal functions $z(t)$ and their cosinereference phasor-domain counterparts $\widetilde{Z}$, where $z(t)=\mathfrak{R e}\left[\tilde{Z} e^{j \omega t}\right]$.

| $z(t)$ | $\widetilde{Z}$ |
| :---: | :---: |
| $A \cos \omega t$ | $\Leftrightarrow \quad A$ |
| $A \cos \left(\omega t+\phi_{0}\right)$ | $\Leftrightarrow A e^{j \phi_{0}}$ |
| $A \cos \left(\omega t+\beta x+\phi_{0}\right)$ | $\Leftrightarrow A e^{j\left(\beta x+\phi_{0}\right)}$ |
| $A e^{-\alpha x} \cos \left(\omega t+\beta x+\phi_{0}\right)$ | $\Leftrightarrow A e^{-\alpha x} e^{j\left(\beta x+\phi_{0}\right)}$ |
| $A \sin \omega t$ | $\Leftrightarrow A e^{-j \pi / 2}$ |
| $A \sin \left(\omega t+\phi_{0}\right)$ | $\Leftrightarrow A e^{j\left(\phi_{0}-\pi / 2\right)}$ |
| $\frac{d}{d t}(z(t))$ | $\leftrightarrow j \omega \widetilde{Z}$ |
| $\frac{d}{d t}\left[A \cos \left(\omega t+\phi_{0}\right)\right]$ | $\Longleftrightarrow \quad j \omega A e^{j \phi_{0}}$ |
| $\int z(t) d t$ | $\Leftrightarrow \frac{1}{j \omega} \widetilde{Z}$ |
| $\int A \sin \left(\omega t+\phi_{0}\right) d t$ | $\Longleftrightarrow \frac{1}{j \omega} A e^{j\left(\phi_{0}-\pi / 2\right)}$ |

PLEASE TURN THE PAGE....

School of Engineering
BEng (Hons) Electrical \& Electronic Engineering
Semester One Examination 2018/2019
Engineering Electromagnetism
Module no. EEE6012

Summary of vector relations.

|  | Cartesian <br> Coordinates | Cylindrical Coordinates | Spherical <br> Coordinates |
| :---: | :---: | :---: | :---: |
| Coordinate variables | $x, y, z$ | $r, \phi, z$ | $R, \theta, \phi$ |
| Vector representation $\mathbf{A}=$ | $\hat{\mathbf{x}} A_{x}+\hat{\mathbf{y}} A_{y}+\hat{\mathbf{z}} A_{z}$ | $\hat{\mathbf{r}} A_{r}+\hat{\phi} A_{\phi}+\hat{\mathbf{z}} A_{z}$ | $\hat{\mathbf{R}} A_{R}+\hat{\boldsymbol{\theta}} A_{\theta}+\hat{\boldsymbol{\phi}} A_{\phi}$ |
| Magnitude of A $\quad\|\mathbf{A}\|=$ | $\sqrt[+]{A_{x}^{2}+A_{y}^{2}+A_{z}^{2}}$ | $\sqrt[+]{A_{r}^{2}+A_{\phi}^{2}+A_{Z}^{2}}$ | $\sqrt[+]{A_{R}^{2}+A_{\theta}^{2}+A_{\phi}^{2}}$ |
| Position vector $\overrightarrow{O P_{1}}=$ | $\begin{gathered} \hat{\mathbf{x}} x_{1}+\hat{\mathbf{y}} y_{1}+\hat{\mathbf{z}} z_{1}, \\ \text { for } P=\left(x_{1}, y_{1}, z_{1}\right) \end{gathered}$ | $\begin{gathered} \hat{\mathbf{r}} r_{1}+\hat{\mathbf{z}} z_{1}, \\ \text { for } P=\left(r_{1}, \phi_{1}, z_{1}\right) \end{gathered}$ | $\hat{\mathbf{R}} R_{1}$, <br> for $P=\left(R_{1}, \theta_{1}, \phi_{1}\right)$ |
| Base vectors properties | $\begin{gathered} \hat{\mathbf{x}} \cdot \hat{\mathbf{x}}=\hat{\mathbf{y}} \cdot \hat{\mathrm{y}}=\hat{\mathbf{z}} \cdot \hat{\mathbf{z}}=1 \\ \hat{\mathbf{x}} \cdot \hat{\mathrm{y}}=\hat{\mathrm{y}} \cdot \hat{\mathbf{z}}=\hat{\mathbf{z}} \cdot \hat{\mathrm{x}}=0 \\ \hat{\mathbf{x}} \times \hat{\mathrm{y}}=\hat{\mathbf{z}} \\ \hat{\mathbf{y}} \times \hat{\mathbf{z}}=\hat{\mathbf{x}} \\ \hat{\mathbf{z}} \times \hat{\mathbf{y}} \end{gathered}$ | $\begin{aligned} & \hat{\mathbf{r}} \cdot \hat{\mathbf{r}}=\hat{\phi} \cdot \hat{\phi}=\hat{\mathbf{z}} \cdot \hat{\mathbf{z}}=1 \\ & \hat{\mathbf{r}} \cdot \hat{\phi}=\hat{\phi} \cdot \hat{\mathbf{z}}=\hat{\mathbf{z}} \cdot \hat{\mathbf{r}}=0 \\ & \hat{\mathbf{r}} \times \hat{\phi}=\hat{\mathbf{z}} \\ & \hat{\phi} \times \hat{\mathbf{z}}=\hat{\mathbf{r}} \\ & \hat{\mathbf{z}} \times \hat{\mathbf{r}}=\hat{\boldsymbol{\phi}} \end{aligned}$ | $\begin{gathered} \hat{\mathbf{R}} \cdot \hat{\mathbf{R}}=\hat{\boldsymbol{\theta}} \cdot \hat{\boldsymbol{\theta}}=\hat{\boldsymbol{\phi}} \cdot \hat{\boldsymbol{\phi}}=1 \\ \hat{\mathbf{R}} \cdot \hat{\boldsymbol{\theta}}=\hat{\boldsymbol{\theta}} \cdot \hat{\boldsymbol{\phi}}=\hat{\boldsymbol{\phi}} \cdot \hat{\mathbf{R}}=0 \\ \hat{\mathbf{R}} \times \hat{\boldsymbol{\theta}}=\hat{\boldsymbol{\phi}} \\ \hat{\boldsymbol{\theta}} \times \hat{\boldsymbol{\phi}}=\hat{\mathbf{R}} \\ \hat{\boldsymbol{\phi}} \times \hat{\mathbf{R}}=\hat{\boldsymbol{\theta}} \end{gathered}$ |
| Dot product $\mathbf{A} \cdot \mathbf{B}=$ | $A_{x} B_{x}+A_{y} B_{y}+A_{z} B_{z}$ | $A_{r} B_{r}+A_{\phi} B_{\phi}+A_{Z} B_{Z}$ | $A_{R} B_{R}+A_{\theta} B_{\theta}+A_{\phi} B_{\phi}$ |
| Cross product $\mathbf{A} \times \mathbf{B}=$ | $\left\|\begin{array}{ccc}\hat{\mathbf{x}} & \hat{\mathbf{y}} & \hat{\mathbf{z}} \\ A_{x} & A_{y} & A_{Z} \\ B_{x} & B_{y} & B_{Z}\end{array}\right\|$ | $\left\|\begin{array}{ccc}\hat{\mathbf{r}} & \hat{\boldsymbol{\phi}} & \hat{\mathbf{z}} \\ A_{r} & A_{\phi} & A_{Z} \\ B_{r} & B_{\phi} & B_{Z}\end{array}\right\|$ | $\left\|\begin{array}{ccc}\hat{\mathbf{R}} & \hat{\boldsymbol{\theta}} & \hat{\boldsymbol{\phi}} \\ A_{R} & A_{\theta} & A_{\phi} \\ B_{R} & B_{\theta} & B_{\phi}\end{array}\right\|$ |
| Differential length $d \mathbf{l}=$ | $\hat{\mathbf{x}} d x+\hat{\mathbf{y}} d y+\hat{\mathbf{z}} d z$ | $\hat{\mathbf{r}} d r+\hat{\boldsymbol{\phi}} r d \phi+\hat{\mathbf{z}} d z$ | $\hat{\mathbf{R}} d R+\hat{\boldsymbol{\theta}} R d \theta+\hat{\boldsymbol{\phi}} R \sin \theta d \phi$ |
| Differential surface areas | $\begin{aligned} & d \mathbf{s}_{x}=\hat{\mathbf{x}} d y d z \\ & d \mathbf{s}_{y}=\hat{\mathbf{y}} d x d z \\ & d \mathbf{s}_{z}=\hat{\mathbf{z}} d x d y \end{aligned}$ | $\begin{aligned} d \mathbf{s}_{r} & =\hat{\hat{r}} r d \phi d z \\ d \mathbf{s}_{\phi} & =\hat{\phi} d r d z \\ d \mathbf{s}_{z} & =\hat{\mathbf{z}} r d r d \phi \end{aligned}$ | $\begin{aligned} d \mathbf{s}_{R} & =\hat{\mathbf{R}} R^{2} \sin \theta d \theta d \phi \\ d \mathbf{s}_{\theta} & =\hat{\boldsymbol{\theta}} R \sin \theta d R d \phi \\ d \mathbf{s}_{\phi} & =\hat{\boldsymbol{\phi}} R d R d \theta \end{aligned}$ |
| Differential volume $d v=$ | $d x d y d z$ | $r d r d \phi d z$ | $R^{2} \sin \theta d R d \theta d \phi$ |

## School of Engineering

BEng (Hons) Electrical \& Electronic Engineering
Semester One Examination 2018/2019
Engineering Electromagnetism
Module no. EEE6012

Coordinate transformation relations.

| Transformation | Coordinate Variables | Unit Vectors | Vector Components |
| :---: | :---: | :---: | :---: |
| Cartesian to cylindrical | $\begin{aligned} & r=\sqrt[+]{x^{2}+y^{2}} \\ & \phi=\tan ^{-1}(y / x) \\ & z=z \end{aligned}$ | $\begin{aligned} & \hat{\mathbf{r}}=\hat{\mathbf{x}} \cos \phi+\hat{\mathbf{y}} \sin \phi \\ & \hat{\phi}=-\hat{\mathbf{x}} \sin \phi+\hat{\mathbf{y}} \cos \phi \\ & \hat{\mathbf{z}}=\hat{\mathbf{z}} \end{aligned}$ | $\begin{aligned} & A_{r}=A_{x} \cos \phi+A_{y} \sin \phi \\ & A_{\phi}=-A_{x} \sin \phi+A_{y} \cos \phi \\ & A_{z}=A_{z} \end{aligned}$ |
| Cylindrical to Cartesian | $\begin{aligned} & x=r \cos \phi \\ & y=r \sin \phi \\ & z=z \end{aligned}$ | $\begin{aligned} & \hat{\mathbf{x}}=\hat{\mathbf{r}} \cos \phi-\hat{\boldsymbol{\phi}} \sin \phi \\ & \hat{\mathbf{y}}=\hat{\mathbf{r}} \sin \phi+\hat{\boldsymbol{\phi}} \cos \phi \\ & \hat{\mathbf{z}}=\hat{\mathbf{z}} \end{aligned}$ | $\begin{aligned} & A_{x}=A_{r} \cos \phi-A_{\phi} \sin \phi \\ & A_{y}=A_{r} \sin \phi+A_{\phi} \cos \phi \\ & A_{z}=A_{z} \end{aligned}$ |
| Cartesian to spherical | $\begin{aligned} & R=\sqrt[+]{x^{2}+y^{2}+z^{2}} \\ & \theta=\tan ^{-1}\left[\sqrt[+]{x^{2}+y^{2}} / z\right] \\ & \phi=\tan ^{-1}(y / x) \end{aligned}$ | $\begin{aligned} \hat{\mathbf{R}}= & \hat{\mathbf{x}} \sin \theta \cos \phi \\ & \quad+\hat{\mathbf{y}} \sin \theta \sin \phi+\hat{\mathbf{z}} \cos \theta \\ \hat{\boldsymbol{\theta}}= & \hat{\mathbf{x}} \cos \theta \cos \phi \\ & \quad+\hat{\mathbf{y}} \cos \theta \sin \phi-\hat{\mathbf{z}} \sin \theta \\ \hat{\boldsymbol{\phi}}= & -\hat{\mathbf{x}} \sin \phi+\hat{\mathbf{y}} \cos \phi \end{aligned}$ | $\begin{aligned} A_{R}= & A_{x} \sin \theta \cos \phi \\ & +A_{y} \sin \theta \sin \phi+A_{z} \cos \theta \\ A_{\theta}= & A_{x} \cos \theta \cos \phi \\ & +A_{y} \cos \theta \sin \phi-A_{z} \sin \theta \\ A_{\phi}= & -A_{x} \sin \phi+A_{y} \cos \phi \end{aligned}$ |
| Spherical to Cartesian | $\begin{aligned} & x=R \sin \theta \cos \phi \\ & y=R \sin \theta \sin \phi \\ & z=R \cos \theta \end{aligned}$ | $\begin{aligned} \hat{\mathbf{x}}= & \hat{\mathbf{R}} \sin \theta \cos \phi \\ & +\hat{\boldsymbol{\theta}} \cos \theta \cos \phi-\hat{\boldsymbol{\phi}} \sin \phi \\ \hat{\mathbf{y}}= & \hat{\mathbf{R}} \sin \theta \sin \phi \\ \quad & +\hat{\boldsymbol{\theta}} \cos \theta \sin \phi+\hat{\boldsymbol{\phi}} \cos \phi \\ \hat{\mathbf{z}}= & \hat{\mathbf{R}} \cos \theta-\hat{\boldsymbol{\theta}} \sin \theta \end{aligned}$ | $\begin{aligned} A_{x}= & A_{R} \sin \theta \cos \phi \\ & +A_{\theta} \cos \theta \cos \phi-A_{\phi} \sin \phi \\ A_{y}= & A_{R} \sin \theta \sin \phi \\ & +A_{\theta} \cos \theta \sin \phi+A_{\phi} \cos \phi \\ A_{z}= & A_{R} \cos \theta-A_{\theta} \sin \theta \end{aligned}$ |
| Cylindrical to spherical | $\begin{aligned} & R=\sqrt[+]{r^{2}+z^{2}} \\ & \theta=\tan ^{-1}(r / z) \\ & \phi=\phi \end{aligned}$ | $\begin{aligned} & \hat{\mathbf{R}}=\hat{\mathbf{r}} \sin \theta+\hat{\mathbf{z}} \cos \theta \\ & \hat{\boldsymbol{\theta}}=\hat{\mathbf{r}} \cos \theta-\hat{\mathbf{z}} \sin \theta \\ & \hat{\boldsymbol{\phi}}=\hat{\boldsymbol{\phi}} \end{aligned}$ | $\begin{aligned} & A_{R}=A_{r} \sin \theta+A_{Z} \cos \theta \\ & A_{\theta}=A_{r} \cos \theta-A_{Z} \sin \theta \\ & A_{\phi}=A_{\phi} \end{aligned}$ |
| Spherical to cylindrical | $\begin{aligned} & r=R \sin \theta \\ & \phi=\phi \\ & z=R \cos \theta \end{aligned}$ | $\begin{aligned} & \hat{\mathbf{r}}=\hat{\mathbf{R}} \sin \theta+\hat{\boldsymbol{\theta}} \cos \theta \\ & \hat{\boldsymbol{\phi}}=\hat{\boldsymbol{\phi}} \\ & \hat{\mathbf{z}}=\hat{\mathbf{R}} \cos \theta-\hat{\boldsymbol{\theta}} \sin \theta \end{aligned}$ | $\begin{aligned} & A_{r}=A_{R} \sin \theta+A_{\theta} \cos \theta \\ & A_{\phi}=A_{\phi} \\ & A_{Z}=A_{R} \cos \theta-A_{\theta} \sin \theta \end{aligned}$ |

## ELECTROSTATICS:

$\mathbf{F}_{12}=\frac{Q_{1} Q_{2}}{4 \pi \varepsilon_{0} R^{2}} \mathbf{a}_{R_{12}}, \mathbf{F}=\frac{Q}{4 \pi \varepsilon_{0}} \sum_{k=1}^{N} \frac{Q_{k}\left(\mathbf{r}-\mathbf{r}_{k}\right)}{\left|\mathbf{r}-\mathbf{r}_{k}\right|^{3}}, \mathbf{E}=\frac{\mathbf{F}}{Q}, \mathbf{E}=\int \frac{\rho_{L} d l}{4 \pi \varepsilon_{0} R^{2}} \mathbf{a}_{R}, \mathbf{E}=\int \frac{\rho_{S} d S}{4 \pi \varepsilon_{0} R^{2}} \mathbf{a}_{R}, \mathbf{E}=\int \frac{\rho_{v} d v}{4 \pi \varepsilon_{0} R^{2}} \mathbf{a}_{R}$
$\mathbf{E}=\frac{\rho_{S}}{2 \varepsilon_{0}} \mathbf{a}_{n}, \mathbf{E}=\frac{\rho_{L}}{2 \pi \varepsilon_{0} \rho} \mathbf{a}_{\rho}, Q=\oint_{S} \mathbf{D} \cdot d \mathbf{S}=\int_{v} \rho_{v} d v, \nabla \cdot \mathbf{D}=\rho_{\nu}, W=-Q \int_{A}^{B} \mathbf{E} \cdot d \ell, V_{A B}=\frac{W}{Q}=-\int_{A}^{B} \mathbf{E} \cdot d \ell, V=\frac{Q}{4 \pi \varepsilon_{0} r}$ $\oint \mathbf{E} \cdot d \ell=0, \nabla \times \mathbf{E}=0, \mathbf{E}=-\nabla V, W_{E}=\frac{1}{2} \sum_{k=1}^{n} Q_{k} V_{k}, W_{E}=\frac{1}{2} \int \mathbf{D} \cdot \mathbf{E} d v=\frac{1}{2} \int \varepsilon_{0} E^{2} d v, \mathbf{J}=\rho_{v} \mathbf{u}, I=\int_{S} \mathbf{J} \cdot d \mathbf{S}, \mathbf{J}=\sigma \mathbf{E}$,
$R=\frac{\mathrm{V}}{\mathrm{I}}=\frac{\int \mathbf{E} \cdot d \mathbf{I}}{\int \sigma \mathbf{E} \cdot d \mathbf{S}}, \mathbf{D}=\varepsilon \mathbf{E}, \nabla \cdot \mathbf{J}=-\frac{\partial \rho_{v}}{\partial t}, E_{1 t}=E_{2 t}, D_{1 n}-D_{2 n}=\rho_{S}, D_{1 n}=D_{2 n}, \frac{\tan \theta_{1}}{\tan \theta_{2}}=\frac{\varepsilon_{r 1}}{\varepsilon_{r 2}}$
$\nabla^{2} V=-\frac{\rho_{V}}{\varepsilon}, \nabla^{2} V=0, C=\frac{Q}{V}=\frac{\varepsilon \oint \mathbf{E} \cdot d \mathbf{S}}{\int \mathbf{E} \cdot d \mathbf{I}}, W_{E}=\frac{1}{2} C V^{2}=\frac{1}{2} Q V=\frac{Q^{2}}{2 C}, C=\frac{Q}{V}=\frac{2 \pi \varepsilon L}{\ln \frac{b}{a}}, C=\frac{Q}{V}=\frac{4 \pi \varepsilon}{\frac{1}{a}-\frac{1}{b}}, R C=\frac{\varepsilon}{\sigma}$

$$
\epsilon_{o}=8.85 \times 10^{-12} \mathrm{~F} / \mathrm{m} \quad, \mu_{o}=4 \pi X 10^{-7} \mathrm{H} / \mathrm{m}
$$

## School of Engineering

BEng (Hons) Electrical \& Electronic Engineering
Semester One Examination 2018/2019
Engineering Electromagnetism
Module no. EEE6012

MAGNETOSTATICS:
$\mathbf{H}=\int_{L} \frac{I d \mathbf{I} \times \mathbf{a}_{R}}{4 \pi R^{2}}, \mathbf{H}=\int_{S} \frac{\mathbf{K} d S \times \mathbf{a}_{R}}{4 \pi R^{2}}, \mathbf{H}=\int_{v} \frac{\mathbf{J} d v \times \mathbf{a}_{R}}{4 \pi R^{2}}, \mathbf{H}=\frac{I}{4 \pi \rho}\left(\cos \alpha_{2}-\cos \alpha_{1}\right) \mathbf{a}_{\phi}, \mathbf{H}=\frac{I}{2 \pi \rho} \mathbf{a}_{\phi}, \mathbf{a}_{\phi}=\mathbf{a}_{\ell} \times \mathbf{a}_{\rho}$,
$\oint \mathbf{H} \cdot d \mathbf{I}=I_{e n c}, \nabla \times \mathbf{H}=\mathbf{J}, \mathbf{H}=\frac{I}{2 \pi \rho} \mathbf{a}_{\phi}, \mathbf{H}=\frac{1}{2} \mathbf{K} \times \mathbf{a}_{n}, \mathbf{B}=\mu \mathbf{H}, \Psi=\int_{S} \mathbf{B} \cdot d \mathbf{S}, \oint \mathbf{B} \cdot d \mathbf{S}=0, \nabla \cdot \mathbf{B}=0, \mathbf{H}=-\nabla \mathrm{V}_{m}$, $\mathbf{B}=\nabla \times \mathbf{A}, \mathbf{A}=\int_{L} \frac{\mu_{0} I d \mathbf{I}}{4 \pi R}, \mathbf{A}=\int_{S} \frac{\mu_{0} \mathbf{K} d S}{4 \pi R}, \mathbf{A}=\int_{v} \frac{\mu_{0} \mathbf{J} d v}{4 \pi R}, \Psi=\oint_{L} \mathbf{A} \cdot d \mathbf{I}, \mathbf{F}=Q(\mathbf{E}+\mathbf{u} \times \mathbf{B}), d \mathbf{F}=I d \mathbf{I} \times \mathbf{B}, \mathbf{B}_{1 n}=\mathbf{B}_{2 n}$, $\left(\mathbf{H}_{1}-\mathbf{H}_{2}\right) \times \mathbf{a}_{n 12}=\mathbf{K}, \mathbf{H}_{1 t}=\mathbf{H}_{2 t}, \frac{\tan \theta_{1}}{\tan \theta_{2}}=\frac{\mu_{1}}{\mu_{2}}, L=\frac{\lambda}{I}=\frac{N \psi}{I}, M_{12}=\frac{\lambda_{12}}{I_{2}}=\frac{N_{1} \psi_{12}}{I_{2}}, W_{m}=\frac{1}{2} \int \mathbf{B} \cdot \mathbf{H} d v=\frac{1}{2} \int \mu H^{2} d v$

## WAVES AND APPLICATIONS

$$
S=\frac{\left|V_{\max }\right|}{\left|V_{\min }\right|}=\frac{1+|\Gamma|}{1-|\Gamma|}
$$

$$
\Gamma=\frac{Z_{L}-Z_{0}}{Z_{L}+Z_{0}}
$$

$$
\begin{aligned}
& \mathrm{V}_{e m f}=-\frac{d \psi}{d t} \quad, \mathrm{~V}_{e m f}=\oint_{L} \mathbf{E} \cdot d \mathbf{I}=-\int_{S} \frac{\partial \mathbf{B}}{\partial t} \cdot d \mathbf{S} \quad, \nabla \times \mathbf{E}=-\frac{\partial \mathbf{B}}{\partial t} \quad, \mathrm{~V}_{e m f}=\oint_{L} \mathbf{E}_{m} \cdot d \mathbf{I}=\oint_{L}(\mathbf{u} \times \mathbf{B}) \cdot d \mathbf{I} \\
& \mathrm{~V}_{e m f}=\oint_{L} \mathbf{E} \cdot d \mathbf{I}=-\int_{S} \frac{\partial \mathbf{B}}{\partial t} \cdot d \mathbf{S}+\oint_{L}(\mathbf{u} \times \mathbf{B}) \cdot d \mathbf{I} \quad, \mathbf{J}_{d}=\frac{\partial \mathbf{D}}{d t} \quad, \nabla \times \mathbf{H}=\mathbf{J}+\frac{\partial \mathbf{D}}{d t} \quad, \beta=\frac{2 \pi}{\lambda}, \underline{\gamma}=\alpha+j \beta \\
& \alpha=\omega \sqrt{\frac{\mu \varepsilon}{2}\left[\sqrt{1+\left[\frac{\sigma}{\omega \varepsilon}\right]^{2}}-1\right]}, \quad \beta=\omega \sqrt{\frac{\mu \varepsilon}{2}\left[\sqrt{1+\left[\frac{\sigma}{\omega \varepsilon}\right]^{2}}+1\right]}, \mathbf{E}(z, t)=E_{0} e^{-\alpha z} \cos \left(\omega t-\beta_{z}\right) \mathbf{a}_{x} \\
& |\underline{\eta}|=\frac{\sqrt{\mu / \varepsilon}}{\left[1+\left(\frac{\sigma}{\omega \varepsilon}\right)^{2}\right]^{1 / 4}}, \quad \tan 2 \theta_{\eta}=\frac{\sigma}{\omega \varepsilon}, \mathbf{H}=\frac{E_{0}}{|\underline{\eta}|} e^{-\alpha c} \cos \left(\omega t-\beta_{z}-\theta_{\eta}\right) \mathbf{a}_{y}, \tan \theta=\frac{\sigma}{\omega \varepsilon}, \mathbf{a}_{E} \times \mathbf{a}_{H}=\mathbf{a}_{k} \\
& \eta_{0}=\sqrt{\frac{\mu_{0}}{\varepsilon_{0}}}=120 \pi \approx 377 \Omega, p(t)=\mathbf{E} \times \mathbf{H}, p_{\text {ave }}(z)=\frac{1}{2} \operatorname{Re}\left(\mathbf{E}_{s} \times \mathbf{H}^{*}\right), p_{\text {ave }}(z)=\frac{E_{0}^{2}}{2|\underline{\eta}|} e^{-2 \alpha c} \cos \theta_{\eta} \mathbf{a}_{z}, P_{\text {ave }}=\int_{S} p_{\text {ave }} \cdot d \mathbf{S}, \\
& \Gamma=\frac{E_{r o}}{E_{\text {io }}}=\frac{\eta_{2}-\eta_{1}}{\eta_{2}+\eta_{1}}, \tau=\frac{E_{\text {to }}}{E_{\text {io }}}=\frac{2 \eta_{2}}{\eta_{2}+\eta_{1}}, s=\frac{\left|\mathbf{E}_{1}\right|_{\max }}{\left|\mathbf{E}_{1}\right|_{\min }}=\frac{\left|\mathbf{H}_{1}\right|_{\max }}{\left|\mathbf{H}_{1}\right|_{\min }}=\frac{1+|\Gamma|}{1-|\Gamma|}, \quad k_{i} \sin \theta_{i}=k_{t} \sin \theta_{t}, \\
& \Gamma_{\|}=\frac{E_{r o}}{E_{\text {io }}}=\frac{\eta_{2} \cos \theta_{t}-\eta_{1} \cos \theta_{i}}{\eta_{2} \cos \theta_{t}+\eta_{1} \cos \theta_{i}}, \tau_{\|}=\frac{E_{\text {to }}}{E_{\text {io }}}=\frac{2 \eta_{2} \cos \theta_{i}}{\eta_{2} \cos \theta_{t}+\eta_{1} \cos \theta_{i}}, \sin ^{2} \theta_{B \|}=\frac{1-\mu_{2} \varepsilon_{1} / \mu_{1} \varepsilon_{2}}{1-\left(\varepsilon_{1} / \varepsilon_{2}\right)^{2}}, \\
& \Gamma_{\perp}=\frac{E_{r o}}{E_{\text {io }}}=\frac{\eta_{2} \cos \theta_{i}-\eta_{1} \cos \theta_{t}}{\eta_{2} \cos \theta_{i}+\eta_{1} \cos \theta_{t}}, \tau_{\perp}=\frac{E_{\text {to }}}{E_{\text {io }}}=\frac{2 \eta_{2} \cos \theta_{i}}{\eta_{2} \cos \theta_{i}+\eta_{1} \cos \theta_{t}}, \sin ^{2} \theta_{B \perp}=\frac{1-\mu_{1} \varepsilon_{2} / \mu_{2} \varepsilon_{1}}{1-\left(\mu_{1} / \mu_{2}\right)^{2}} \\
& \omega=\beta c
\end{aligned}
$$

School of Engineering
BEng (Hons) Electrical \& Electronic Engineering
Semester One Examination 2018/2019
Engineering Electromagnetism
Module no. EEE6012

## Antenna and Radar formula

## Dipole

Solid angle:

$$
\Omega_{\mathrm{p}}=\iint_{4 \pi} F(\theta, \phi) d \Omega
$$

Directivity:

$$
D=\frac{4 \pi}{\Omega_{\mathrm{p}}} \quad D=\frac{4 \pi A_{\mathrm{e}}}{\lambda^{2}}
$$

## Shorted dipole

$$
\begin{aligned}
& S_{0}=\frac{15 \pi I_{0}^{2}}{R^{2}}\left(\frac{l}{\lambda}\right)^{2} \\
& R_{\mathrm{rad}}=80 \pi^{2}\left(\frac{l}{\lambda}\right)^{2} .
\end{aligned}
$$

Hertzian monopole

$$
\begin{aligned}
& R_{\mathrm{rad}}=80 \pi^{2}\left[\frac{\mathrm{~d} l}{\lambda}\right]^{2} \\
& P_{\mathrm{rad}}=\frac{1}{2} I_{\mathrm{o}}^{2} R_{\mathrm{rad}}
\end{aligned}
$$

## School of Engineering

## BEng (Hons) Electrical \& Electronic Engineering

Semester One Examination 2018/2019
Engineering Electromagnetism
Module no. EEE6012

## Half -wave dipole

$\widetilde{E}_{\theta}=j 60 I_{0}\left\{\frac{\cos [(\pi / 2) \cos \theta]}{\sin \theta}\right\}\left(\frac{e^{-j k R}}{R}\right)$,
$\widetilde{H}_{\phi}=\frac{\widetilde{E}_{\theta}}{\eta_{0}}$.
$\left|E_{\phi s}\right|=\frac{\eta_{\mathrm{o}} I_{\mathrm{o}} \cos \left(\frac{\pi}{2} \cos \theta\right)}{2 \pi r \sin \theta}$
$\left|H_{\phi s}\right|=\frac{I_{\mathrm{o}} \cos \left(\frac{\pi}{2} \cos \theta\right)}{2 \pi r \sin \theta}$

For Transmission line

|  | Propagation Constant $\gamma=\alpha+j \beta$ | Phase Velocity $u_{\mathrm{p}}$ | Characteristic Impedance $Z_{0}$ |
| :---: | :---: | :---: | :---: |
| General case | $\gamma=\sqrt{\left(R^{\prime}+j \omega L^{\prime}\right)\left(G^{\prime}+j \omega C^{\prime}\right)}$ | $u_{\mathrm{p}}=\omega / \beta$ | $Z_{0}=\sqrt{\frac{\left(R^{\prime}+j \omega L^{\prime}\right)}{\left(G^{\prime}+j \omega C^{\prime}\right)}}$ |
| Lossless $\left(R^{\prime}=G^{\prime}=0\right)$ | $\alpha=0, \beta=\omega \sqrt{\varepsilon_{\mathrm{r}}} / c$ | $u_{\mathrm{p}}=c / \sqrt{\varepsilon_{\mathrm{r}}}$ | $Z_{0}=\sqrt{L^{\prime} / C^{\prime}}$ |
| Lossless coaxial | $\alpha=0, \beta=\omega \sqrt{\varepsilon_{\mathrm{r}}} / c$ | $u_{\mathrm{p}}=c / \sqrt{\varepsilon_{\mathrm{r}}}$ | $Z_{0}=\left(60 / \sqrt{\varepsilon_{\mathrm{r}}}\right) \ln (b / a)$ |
| Lossless two-wire | $\alpha=0, \beta=\omega \sqrt{\varepsilon_{\mathrm{r}}} / c$ | $u_{\mathrm{p}}=c / \sqrt{\varepsilon_{\mathrm{r}}}$ | $\begin{aligned} Z_{0}= & \left(120 / \sqrt{\varepsilon_{\mathrm{r}}}\right) \\ & \cdot \ln \left[(D / d)+\sqrt{(D / d)^{2}-1}\right] \end{aligned}$ |
|  |  |  | $\begin{aligned} & Z_{0} \simeq\left(120 / \sqrt{\varepsilon_{\mathrm{r}}}\right) \ln (2 D / d), \\ & \text { if } D \gg d \end{aligned}$ |
| Lossless parallel-plate | $\alpha=0, \beta=\omega \sqrt{\varepsilon_{\mathrm{r}}} / c$ | $u_{\mathrm{p}}=c / \sqrt{\varepsilon_{\mathrm{r}}}$ | $Z_{0}=\left(120 \pi / \sqrt{\varepsilon_{\mathrm{r}}}\right)(h / w)$ |

Notes: (1) $\mu=\mu_{0}, \quad \varepsilon=\varepsilon_{\mathrm{r}} \varepsilon_{0}, c=1 / \sqrt{\mu_{0} \varepsilon_{0}}$, and $\sqrt{\mu_{0} / \varepsilon_{0}} \simeq(120 \pi) \Omega$, where $\varepsilon_{\mathrm{r}}$ is the relative permittivity of insulating material. (2) For coaxial line, $a$ and $b$ are radii of inner and outer conductors. (3) For two-wire line, $d=$ wire diameter and $D=$ separation between wire centers. (4) For parallel-plate line, $w=$ width of plate and $h=$ separation between the plates.

School of Engineering
BEng (Hons) Electrical \& Electronic Engineering
Semester One Examination 2018/2019
Engineering Electromagnetism
Module no. EEE6012
Distortionless line
$\gamma=\sqrt{R G}+\mathrm{j} \omega \sqrt{L C}$
$\frac{R}{L}=\frac{G}{C}, \quad Z_{o}=\sqrt{\frac{L}{C}}$
Open-circuited line
$\widetilde{V}_{\mathrm{OC}}(d)=V_{0}^{+}\left[e^{j \beta d}+e^{-j \beta d}\right]=2 V_{0}^{+} \cos \beta d$,
$\tilde{I}_{\mathrm{oc}}(d)=\frac{V_{0}^{+}}{Z_{0}}\left[e^{j \beta d}-e^{-j \beta d}\right]=\frac{2 j V_{0}^{+}}{Z_{0}} \sin \beta d$,
$Z_{\mathrm{in}}^{\mathrm{oc}}=\frac{\widetilde{V}_{\mathrm{Oc}}(l)}{\tilde{I}_{\mathrm{oc}}(l)}=-j Z_{0} \cot \beta l$.

Short-circuited line

$$
\begin{aligned}
& \widetilde{V}_{\mathrm{sc}}(d)=V_{0}^{+}\left[e^{j \beta d}-e^{-j \beta d}\right]=2 j V_{0}^{+} \sin \beta d, \\
& \tilde{I}_{\mathrm{sc}}(d)=\frac{V_{0}^{+}}{Z_{0}}\left[e^{j \beta d}+e^{-j \beta d}\right]=\frac{2 V_{0}^{+}}{Z_{0}} \cos \beta d, \\
& Z_{\mathrm{sc}}(d)=\frac{\widetilde{V}_{\mathrm{sc}}(d)}{\widetilde{I}_{\mathrm{sc}}(d)}=j Z_{0} \tan \beta d . \\
& j \omega L_{\mathrm{eq}}=j Z_{0} \tan \beta l, \quad \text { if } \tan \beta l \geq 0 \\
& \frac{1}{j \omega C_{\mathrm{eq}}}=j Z_{0} \tan \beta l, \quad \text { if } \tan \beta l \leq 0
\end{aligned}
$$

$$
Z_{\text {in }}=Z_{\mathrm{o}}\left[\frac{Z_{L}+j Z_{0} \tan \beta \ell}{Z_{\mathrm{o}}+j Z_{L} \tan \beta \ell}\right]
$$

$$
Z_{\mathrm{in}}=Z_{\mathrm{o}}\left[\frac{Z_{L}+Z_{\mathrm{o}} \tanh \gamma \ell}{Z_{\mathrm{o}}+Z_{L} \tanh \gamma \ell}\right]
$$

School of Engineering
BEng (Hons) Electrical \& Electronic Engineering
Semester One Examination 2018/2019
Engineering Electromagnetism
Module no. EEE6012
$V_{\mathrm{o}}=\frac{Z_{\text {in }}}{Z_{\text {in }}+Z_{g}} V_{g} \quad I_{\mathrm{o}}=\frac{V_{g}}{Z_{\text {in }}+Z_{g}}$
$V_{o}=V_{t} e^{j \beta t}$
For a bistatic radar (one in which the transmitting and receiving antennas are separated), the power received is given by

$$
P_{r}=\frac{G_{d t} G_{d r}}{4 \pi}\left[\frac{\lambda}{4 \pi r_{1} r_{2}}\right]^{2} \sigma P_{\mathrm{rad}}
$$

For a monostatic radar, $r_{1}=r_{2}=r$ and $G_{d t}=G_{d r}$.

$$
P_{\mathrm{rec}}=P_{\mathrm{t}} G_{\mathrm{t}} G_{\mathrm{r}}\left(\frac{\lambda}{4 \pi R}\right)^{2}
$$

