## **UNIVERSITY OF BOLTON**

## **SCHOOL OF ENGINEERING**

## **BENG (HONS) IN BIOMEDICAL ENGINEERING**

## **SEMESTER ONE EXAMINATION 2018/2019**

## **ADVANCED BIOMECHATRONIC SYSTEMS**

# MODULE NO: BME6003

Date: Friday 18 January 2019

Time: 10.00 am-12.00 noon

**INSTRUCTIONS TO CANDIDATES:** 

There are <u>SIX</u> questions. You are required to answer <u>ANY FOUR</u> questions.

All questions carry equal marks.

Marks for parts of questions are shown in brackets.

**CANDIDATES REQUIRE:** 

Property tables provided Formula Sheet (attached) Take density of water as 1000 kg/m<sup>3</sup>

## Q1

(a) The following experimental data presented in Table Q1(a) were recorded by using a sinusoidal signal as the input and monitoring the steady-state output to determine the Gain and the Phase shift between the output and input.

$able \mathbf{w} = (a) (a) (a) (a) (a) (a) (b) (a) (b) (b) (b) (b) (b) (b) (b) (b) (b) (b$				
Output	Phase Shift			
(Volts peak-peak)	(Degree)			
0.4	86			
1.2	78			
1.5	58			
1.9	46			
3.8	32			
4.6	24			
5.8	12			
	Output           (Volts peak-peak)           0.4           1.2           1.5           1.9           3.8           4.6			

#### Table Q1 (a) Input (Volts peak-peak) = 6 (V)

Analyse these data from the table and answer:

i)	How to determine the Gain.	[2 marks]
ii)	What the meaning of Phase shift is.	[2 marks]
iii)	What kind of system it is.	[2 marks]

(b) Explain, helped by sketches, what frequency response is and why it is useful for biomechatronic systems control.

[6 marks]

(c) Figure Q1(c) (on the following page) shows a Bode plot.

- i) Estimate the gain margin and the phase margin.
- ii) Explain the functions of gain margin and phase margin in systems control.
- iii) Comment on the system's stability performance.

[2 marks]

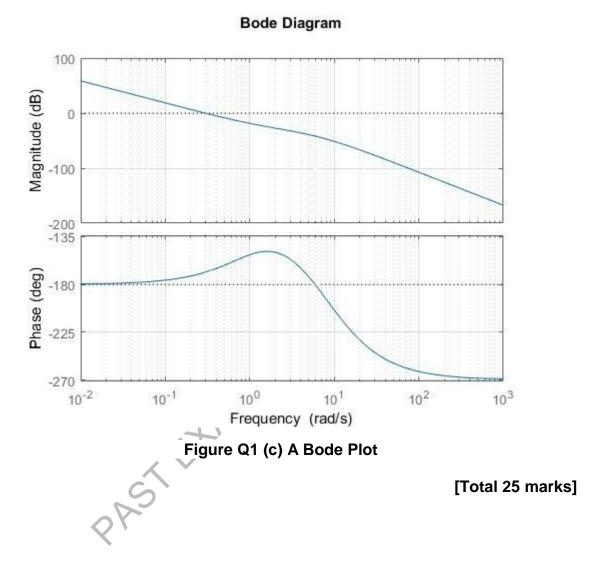
iv) Explain the system's Peak Resonance  $M_p$  and Bandwidth.

[3 marks]

[4 marks]

Q1 continues over the page..... PLEASE TURN THE PAGE.....

#### Q1 continued.....



## Q2

A simplified model of a prosthesis arm system is shown in Figure Q2. The control system for the prosthesis limb dynamics is given by:

$$G_p(s) = \frac{8}{(2s+4)(s+2)}$$

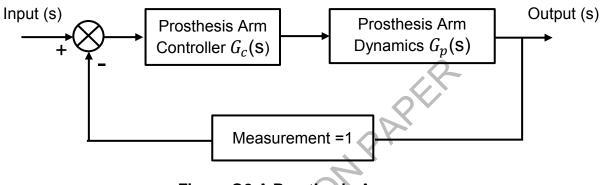


Figure Q2 A Prosthesis Arm

The system responses for a unit step input are required as:

- The maximum overshoot is less than 12%
- The rise time is less than 0.4 seconds
- The steady-state error is 0\_
- (a) If Gc (s) is a PID controller with Kp = 2, Ki = 5 and Kd = 3, find the range of the gain Kp making the system to be an underdamped system for unit step input. examine the actual system's percent overshoot, rise time and steadystate error under the PID controller and check whether the design criteria have been achieved or not.

## [14 marks]

(b) Describe, helped by sketches, how the error item is handled by proportional, integral and derivative controller.

## [8 marks]

(c) If the design criteria haven't been achieved by using the PID controller provided above, explain the procedure to modify the PID controller.

[3 marks]

[Total 25 marks]

## Q3

- (a) Using block diagrams, briefly explain an analogue closed loop control system and a digital closed loop control system. Assume that input, output and sensor signals for both control systems are all analogue signals. [6 marks]
- (b) Explain what is meant by a zero-order hold (ZOH) system.

[4 marks]

(c) A controller has an 12 bit Analogue to Digital Converter with the signal range between 0 Volt to +12 Volt:

(i)	What is the resolution of the AD converter?	
(ii)	What integer number represented a value of +6 Volts?	[2 marks]
		[2 marks]
(iii)	What voltage does the integer 3072 represent?	[2 marks]
(iv)	What voltage does 010011100101 represent?	
		[2 marks]

(d) A controller of biomechatronic system consists of a Digital to Analogue Converter with zero order element in series with the processing centre which has a transfer function

$$G_p(s) = \frac{6}{s(s+6)}$$

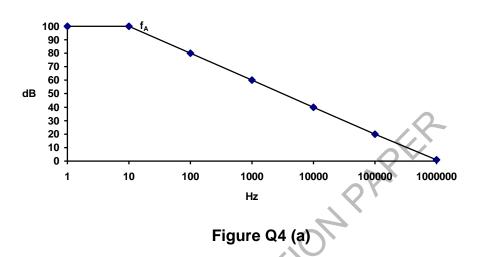
Find the sampled-data transfer function, G (z) for the digital control system. The sampling time, T, is 1 seconds.

[7 marks] [Total 25 marks]

#### Q4

(a)A typical op-amp has an open loop gain graph shown in Figure Q4 (a), derive a transfer function which predicts the open loop gain/frequency response.

[5 marks]



(b)If a non-inverting amplifier configuration has a closed loop gain of eleven, sketch a suitable circuit and calculate the bandwidth and input/output resistance of the non-inverting amplifier using the transfer function from part (a), if Rin is  $100k\Omega$  and Rout is  $100\Omega$ .

## [10 marks]

(c)A low pass first order filter is connected to the input of the amplifier (in part b), sketch the frequency/phase plots of the modified circuit, if the low pass filter uses a resistor of  $100k\Omega$  and a capacitor of 100nF.

[10 marks] [Total 25 marks]

## Q5

(a)When describing an OP-Amp, what is meant by the terms; common - mode rejection ratio, slew rate and transition frequency.

[6 marks]

(b)For the circuit shown in Figure Q5 (b) calculate the output voltage.

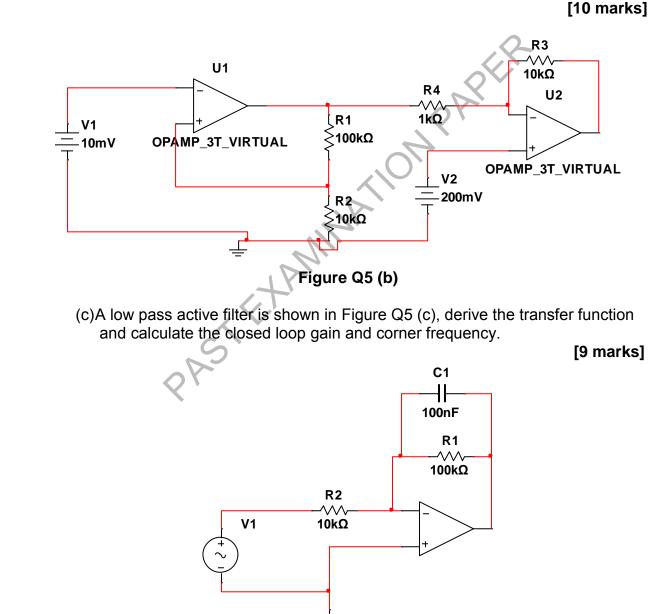


Figure Q5 (c)

[Total 25 marks]

### Q6

a) A Biomedical system has the following closed loop transfer function; sketch the magnitude and phase plots and the final closed loop plot.

[10 marks]

$$T_{(s)} = \frac{10(0.5\,s+1)}{(0.2+1)}$$

b) Describe what is meant by gain and phase margins.

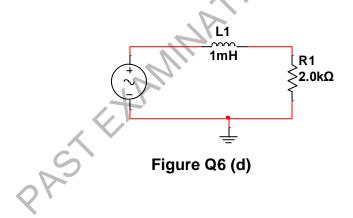
### [2 marks]

c) A low pass filter is shown in Figure Q4, derive the transfer function and sketch magnitude and phase.

### [6 marks]

d) By deriving the transfer function for the circuit shown in Figure Q6 (d) and converting to the s-domain, sketch the pole zero diagram and indicate on the diagram all relevant points.

#### [7 marks]



[Total 25 marks]

**END OF QUESTIONS** 

#### **Formula Sheets**

#### Blocks with feedback loop

$$G(s) = \frac{G_o(s)}{1 + G_o(s)H(s)}$$
 (for a negative feedback)

$$G(s) = \frac{G_o(s)}{1 - G_o(s)H(s)}$$
 (for a negative feedback)

#### Steady-State Errors

$$\begin{split} e_{ss} &= \lim_{s \to 0} [s(1 - G_o(s))\theta_i(s)] \text{ (for an open - loop system)} \\ e_{ss} &= \lim_{s \to 0} \left[ s \frac{1}{1 + G_o(s)} \theta_i(s) \right] \text{ (for the closed - loop system with a unity feedback)} \\ e_{ss} &= \lim_{s \to 0} \left[ s \frac{1}{1 + \frac{G_1(s)}{1 + G_1(s)[H(s) - 1]}} \theta_i(s) \right] \text{ (if the feedback H(s) $\neq 1$)} \\ e_{ss} &= \lim_{s \to 0} \left[ -s \frac{G_2(s)}{1 + G_2(G_1(s) + 1)} \theta_d(s) \right] \text{ (if the system subjects to a disturbance input)} \end{split}$$

#### **First order Systems**

$$G(s) = \frac{\theta_o}{\theta_i} = \frac{G_{ss}(s)}{\tau s + 1}$$
  

$$\tau \left(\frac{d\theta_o}{dt}\right) + \theta_o = G_{ss}\theta_i$$
  

$$\theta_o = G_{ss} \left(1 - e^{-t/\tau}\right) \text{ (for a unit step input)}$$
  

$$\theta_o = AG_{ss} \left(1 - e^{-t/\tau}\right) \text{ (for a step input with size A)}$$
  

$$\theta_o = G_{ss} \left(\frac{1}{\tau}\right) e^{-(t/\tau)} \text{ (for an impulse input)}$$

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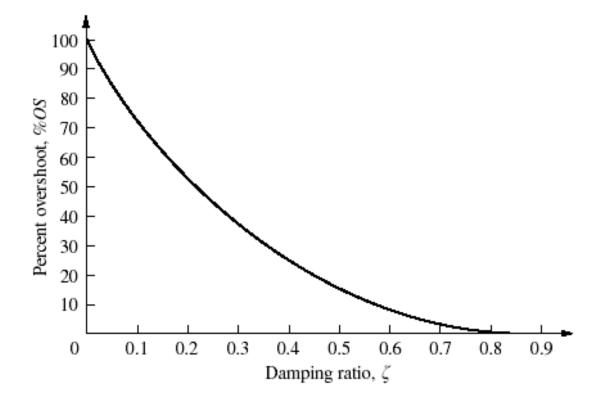
## Second- order Systems

$$\frac{d^2\theta_o}{dt^2} + 2\zeta\omega_n\frac{d\theta_o}{dt} + \omega_n^2\theta_o = b_o\omega_n^2\theta_i$$
$$G(s) = \frac{\theta_o(s)}{\theta_i(s)} = \frac{b_o\omega_n^2}{s^2 + 2\zeta\omega_n s + \omega_n^2}$$

$$\omega_d t_r = 1/2\pi \qquad \qquad \omega_d t_p = \pi$$

p.o.= exp
$$\left(\frac{-\zeta \pi}{\sqrt{(1-\zeta^2)}}\right)$$
 × 100%

$$t_s = \frac{4}{\zeta \omega_n} \qquad \qquad \omega_d = \omega_n \sqrt{(1 - \zeta^2)}$$



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Laplace transform	Time function	Description of time function
1		A unit impulse
1.		A unit step function
$\frac{e^{-st}}{s}$	· · ·	A delayed unit step function
$\frac{1-e^{-st}}{s}$	· · · · ·	A rectangular pulse of duration 7
$\frac{1}{s^2}$	<i>t</i> ,	A unit slope ramp function
$\frac{1}{s^3}$	$\frac{t^2}{2}$	
$\frac{1}{s+a}$	e <sup>-at</sup>	Exponential decay
$\frac{1}{(s+a)^2}$	$te^{-at}$	
$\frac{2}{(s+a)^3}$	$t^2 e^{-at}$	
$\frac{a}{s(s+a)}$	$1 - e^{-at}$	Exponential growth
$\frac{a}{s^2(s+a)}$	$t = \frac{(1 - e^{-at})}{a}$	1
$\frac{a^2}{s(s+a)^2}$	$1 - e^{-at} - ate^{-at}$	
$\frac{s}{(s+a)^2}$	$(1-at)e^{-at}$	
$\frac{1}{(s+a)(s+b)}$	$\frac{e^{-at} - e^{-bt}}{b - a}$	
$\frac{ab}{s(s+a)(s+b)}$	$1 - \frac{b}{b-a}e^{-at} + \frac{a}{b-a}e^{-bt}$	· · · ·
$\frac{1}{(s+a)(s+b)(s+c)}$	$\frac{e^{-at}}{(b-a)(c-a)} + \frac{e^{-bt}}{(c-a)(a-b)} + \frac{e^{-ct}}{(a-c)(b-c)}$	
$\frac{\omega}{s^2 + \omega^2}$	$\sin \omega t$ .	Sine wave
$\frac{s}{s^2 + \omega^2}$	cos wt	Cosine wave
$\frac{\omega}{(s+a)^2+\omega^2}$	$e^{-at} \sin \omega t$	Damped sine wave
$\frac{s+a}{(s+a)^2+\omega^2}$	$e^{-at} \cos \omega t$	Damped cosine wave
$\frac{\omega^2}{s(s^2+\omega^2)}$	$1 - \cos \omega t$	
$\frac{\omega^2}{s^2 + 2\zeta\omega s + \omega^2}$	$\frac{\omega}{\sqrt{(1-\zeta^2)}}e^{-\zeta\omega t}\sin\left[\omega\sqrt{(1-\zeta^2)t}\right]$	4 . <b>.</b> .
$\frac{\omega^2}{s(s^2+2\zeta\omega s+\omega^2)}$	$1 - \frac{1}{\sqrt{(1-\zeta^2)}} e^{-\zeta \omega t} \sin \left[\omega \sqrt{(1-\zeta^2)t} + \phi\right]$	
with $\zeta < 1$	with $\zeta = \cos \phi$	

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Table 15.1 z-transforms	
Sampled f(t), sampling period T	<i>F</i> ( <i>z</i> )
Unit impulse, $\delta(t)$	1
Unit impulse delayed by $kT$	z-* .
Unit step, $u(t)$	$\frac{z}{z-1}$
Unit step delayed by $kT$	$\frac{z}{z^k(z-1)}$
Unit ramp, t	$\frac{Tz}{(z-1)^2}$
<i>t</i> <sup>2</sup>	$\frac{T^2 z(z+1)}{(z-1)^3}$
e <sup>-at</sup>	$\frac{z}{z-e^{-aT}}$
$1 - e^{-at}$	$\frac{z(1-e^{-aT})}{(z-1)(z-e^{-aT})}$
$t e^{-at}$	$\frac{Tz \ \mathrm{e}^{-aT}}{(z - \mathrm{e}^{-aT})^2}$
$e^{-at} - e^{-bt}$	$\frac{(e^{-aT} - e^{-bT})z}{(z - e^{-aT})(z - e^{-bT})}$
sin <i>wt</i>	$\frac{z\sin\omega T}{z^2 - 2z\cos\omega T + 1}$
$\cos \omega t$	$\frac{z(z-\cos\omega T)}{z^2-2z\cos\omega T+1}$
$e^{-at}\sin\omega t$	$\frac{z e^{-aT} \sin \omega T}{z^2 - 2z e^{-aT} \cos \omega T + e^{-2}}$
$e^{-at}\cos\omega t$	$\frac{z(z - e^{-aT}\cos\omega T)}{z^2 - 2z e^{-aT}\cos\omega T + e^{-2}}$

# Table 15.2 z-transforms

<i>f</i> [ <i>k</i> ]	$f[0], f[1], f[2], f[3], \dots$	<i>F</i> ( <i>z</i> )
1u[k]	1, 1, 1, 1,	$\frac{z}{z-1}$
a <sup>k</sup>	$a^0, a^1, a^2, a^3, \dots$	$\frac{z}{z-a}$
k	0, 1, 2, 3,	$\frac{z}{(z-1)^2}$
ka <sup>k</sup>	$0, a^1, 2a^2, 3a^3, \dots$	$\frac{az}{(z-a)^2}$
ka <sup>k-1</sup>	$0, a^0, 2a^1, 3a^2, \dots$	$\frac{z^2}{(z-a)^2}$
e <sup>-ak</sup>	 e <sup>0</sup> , e <sup>-a</sup> , e <sup>-2a</sup> , e <sup>-3a</sup> ,	$\frac{(z-a)^2}{z-e^{-a}}$

### END OF FORMULAE SHEETS